High-speed Digitization Readout of Silicon Photomultipliers for Time of Flight Positron Emission Tomography

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Abstract. We report on work to develop a system with about 100 picoseconds (ps) time resolution for time of flight positron emission tomography [TOF-PET]. The chosen photo detectors for the study were Silicon Photomultipliers (SiPM's). This study was based on extensive experience in studying timing properties of SiPM's [1-4]. The readout of these devices used the commercial high speed digitizer DRS4 [5]. We applied different algorithms to get the best time resolution of 155 ps Guassian (sigma) for a LYSO crystal coupled to a SiPM. We consider the work as a first step in building a prototype TOF-PET module.

I. Introduction

The field of positron-emission-tomography (PET) has been rapidly developing. But there are significant limitations in how well current PET scanners can reconstruct images, related to how fast data can be acquired, how much volume they can image, and the spatial and temporal resolution of the generated photons. Typical modern scanners now include multiple rings of detectors, which can image

a large volume of the patient. In this type of scanner, one can treat each ring as a separate detector and require coincidences only within the ring, or treat the entire region viewed by the scanner as a single 3 dimensional volume. This 3d technique has significantly better sensitivity since more photon pair trajectories are accepted. However, the scattering of photons within the volume of the patient, and the effect of random coincidences limits the technique. The advent of sub-nanosecond timing resolution detectors means that there is potentially much better rejection of scattered photon events and random coincidence events in the 3D technique. In addition, if the timing is good enough, then the origin of photons pairs can be determined better, resulting in improved spatial resolution - so called 'Time-of-Flight' PET, or TOF-PET. Currently a lot of activity has occurred in applications of SiPMs for TOF-PET. [6-8]. This is due to the devices' very good time resolution, low profile, lack of high voltage needed, and their non-sensitivity to magnetic fields. While investigations into this technique have begun elsewhere, we feel that the SiPM characterization extensive and data acquisition expertise of Fermilab, and the historical in-depth research of PET imaging at University of Chicago will combine to make significant strides in this field. We also benefit by a working relationship with the SiPM producer STMicroelectronics (STM) [3,4].

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II. SiPM Layout, role of clipping capacitance

We present here results obtained from SiPM's produced by STM, as well as devices obtained from Hamamatsu. The shape of the SiPM signal plays an important role for improvement of the time resolution [9]. During an extensive series of timing tests we have performed on SiPM's, we would often use a capacitive coupling (10 pF) to clip the slow tail of the SiPM response (see Figure 1 for a schematic). This was performed due to the input requirements of the Ortec 9327 constant fraction discriminator unit we use. Figure 2 presents the effect of the capacitance as an illustration. The use of a high bandwidth amplifier if needed, was shown not to change the pulse shape. The STM SiPM was illuminated by 405 nm light coming from a high speed PiLas laser. For this light input, the signal was suppressed by a factor of 30 due to the capacitive coupling. For the TOF-PET application the shape of the signal is a significant issue. The sharpening of the signal (including its rise time) due to the coupling should improve the time resolution, but it can worsen the energy resolution. That is why we investigated the influence of the clipping capacitance and preamplifiers on the time and energy resolution. The Pomona boxes with SiPMs, circuit boards and mechanics for radiator (or crystal) support is shown in fig. 3.

III. Pulser and Laser Data

The schematic diagram of the readout with PiLas laser as light source is shown in fig. 4. The DRS4 is a 5 GHz digitizer developed by Stefan Ritt at Paul Scherrer Institute, Switzerland [4], and was used to in the measurements reported here. We used the version of the DRS4 with 4 input channels. The DRS4 allows a digitization of the input with a sampling rate of 5 GS/s with an individual channel depth of 1024. The schematic diagram of the readout with the DRS4 is shown in fig. 8. Sometimes we used Ortec V120C to amplify the SiPM signals to fit the DRS4 dynamic range.

First we measured the DRS4 electrical time resolution. The PiLas NIM output was used as trigger for the DRS4. Another NIM signal was split

with a high bandwidth splitter into two identical signals. Each half was applied to channels 1 and 2 of the DRS4. The linear fit on the leading edge of the signals was applied (fig. 5). The fit value for the time at the signals half maximum was taken and the time difference between these 2 values was analyzed. This very simple algorithm was used as a starting point for our analysis. The time spectrum we obtained demonstrated a Gaussian distribution with 8 picoseconds sigma.

The same procedure was repeated with STM signals instead of NIM pulses. The two STM SiPMs were placed at the same distance from the PiLas laser head and were illuminated by the laser light. The STMs overvoltage was 5 Volts and both were operated under room temperature which was $+25^{\circ}$ C. The PiLas light signal was 33 ps full width at half maximum, the wavelength was 405 nm, most data taken with the repetition rate of 100 Hz. We could change the light intensity by the "TUNE" laser option, also as by applying optical filters between the laser head and the SiPMs. The data were taken with the 10 pF capacitance in place. The pulse height distribution (PH) was obtained by integrating the charge collected over all time bins in each DRS4 sample. A few tens of samples before the signal appearance were taken to estimate the base line of the signal. The measured number of photoelectrons (PE) was about 25 (Figure 6). The number of PE was obtained by measuring the PH distribution and fitting it with a Gaussian. The width of the Gaussian fit will indicate the statistical nature of the PE spread. For this laser data, the time resolution with 100 PE was 18 ps and with 25 PE was 39 ps, which roughly follows the expected inverse square root dependence of the timing on the number of PE. This very simple initial approach revealed that the result we obtained for the laser data is consistent with the earlier NIM pulse results [1]. The time resolution without the 10 pF capacitance, using the leading edge approximation, was about the same. These preliminary results showed that best time resolution corresponds to when there is 60 - 400mV maximum signal amplitude, corresponding to the mid-range of the DRS4 digitizer. We also found that the time resolution for this data is dependent on the algorithm we used. The issue will be considered in more detail below.

IV. PET Setup, with Radioactive Source Data

Figure 7 shows the external view of the experimental TOF-PET setup. It consists of two SiPMs with optically attached crystals inside of Pomona boxes. These devices face a Na-22 radioactive source from opposite sides. A Keithley 2410 power supply used to bias the SiPMs [10]. As with the previous laser and NIM pulse data, the signals from the SiPMs were split into two halves, with one half going to form a coincidence trigger and the other half being digitized by the DRS4. We measured time and pulse height distributions with varying:

- 1. SiPMs: We used STM and Hamamatsu devices. The STM sensitive area is 3.5x3.5 mm², with a pixel count of 3,600, pixel size 58x58 um², photon detection efficiency (PDE) about 18% for blue light, a breakdown voltage ~28 Volts, gain of up to 10^6 , overvoltage of 5 Volts. The Hamamatsu sensitive area is 3x3 mm², pixel count 3,600, pixel size 50x50 um², PDE is about 40% for blue light, breakdown voltage ~70 Volts, gain is up to 10^6 , overvoltage is up to 2.5 Volts.
- 2. Crystals consisting of LYSO of different size. The LYSO crystals sizes were, 3x3x10 mm³, 3x3x15 mm³3x3x20 mm³ and 2x2x7 mm³.
- 3. Clipping capacitance and without it.
- 4. Radioactive source Co60.
- 5. Radioactive source Na22.

We found the shape of the signals does not depend on the crystal size we used. The SiPMs coincidence counting rate was about 30 Hz for both Co60 and Na22 sources. The distance between the 2 counters was minimal to place the sources between them and was slightly increased to accommodate the larger crystals size. We found the larger distance does not change significantly time and pulse height resolution with only going down the coincidence counting rate. Optical grease was used to get a good optical contact between the crystals and SiPMs. The rest of the LYSO sides were covered by a few layers of thin TEFLON tape.

The shape of the signals was strongly dependent on whether we used clipping capacitance or not. The traces of STM and Hamamatsu signals with 10 pF clipping capacitance and without it are shown in fig. 9.

The examples of the time and PH distributions are presented in fig. 10. The best time resolution obtained was 155 ps with events corresponding to the 1.17- 1.33 MeV energy of the Co60 photon energy. A cut in PH distribution was applied to retain only the photoelectric interaction peak.

The conditions were: Hamamatsu SiPM, 2x2x7 mm³ LYSO crystal, VT120C amplifier, Co60 radioactive source.

We performed the same test with Na22 source. The best time resolution obtained was 244 ps for MPPCs and 334 ps for STM with LYSO crystals, 3x3x15 mm³ crystals in both cases. The data are taken with 10 pF of the clipping capacitance. The examples of the time and PH distributions are presented for MPPC in fig. 11, and for STM in fig. 12. MPPC and STM results obtained with the same pulse function approach (see below).

V. Algorithms

We started the data analysis with fitting to the leading edge of the signals by straight line, as we discussed above. The point at half maximum of the signal amplitude was detected and used as the timing point. The spectra of the time difference for the two SiPMs signals by using these points were analyzed first. Another analysis was performed such that we fit where the straight line from the fit crossed the fit signal base line. This approach was applied both to data obtained with the laser and with a radioactive source. As a result of these investigations we found that the natural shape of the signals leading edge is far from a straight line, especially for data with the radioactive source and without any shaping of the signals. Even for short signals we observed different slopes along the leading edge.

This is why a functional form approach was introduced. The function well describes different conditions, e.g. SiPM signals with and without the clipping capacitance. A simple model of a SiPM as charging/discharging capacitance was taken as a first approximation to the pulse function. The analytical expression for this function is:

 $p(t) = (1 - exp(-t/\tau 1)) exp(-t/\tau 2),$

where $\tau 1$, $\tau 2$ are charging and discharging time constants. To take into account experimental conditions (jitter, etc.) we convolute the pulse function with resolution function. We use gaussian resolution function with parameter σ which represents an experimental resolution. We got significant improvement of the calibration time resolution (4 ps instead of 8 ps) by applying the function to the PiLas data. To analyze data obtained with LYSO crystal we convoluted the pulse function with exponent with parameter T which represents crystal decay time. Fit to crystal data gives a realistic value for T of about 40 ns. We got a good fit of the function to the real SiPM signals for both PiLas and radioactive source data. The data from the LYSO crystal show significant fluctuations of the signal shape. We apply two stage approach to the crystal data. We fit pulse function to the whole signal to find position of the leading edge. Then we fix discharge time parameter τ_2 and fit the leading edge region only.

VI. Discussion

Both single photoelectron SiPm signal and crystal light pulses have very sharp leading edge and much slower falling tail. It was a bit surprising for us to see that convolution of the SiPms signal shape with crystal light pulse shape changes dramatically the leading edge of the resulting signal. Our result is turned out to be due to the long falling tail of the SiPms signals in accordance with [5]. Shortening of the SiPM signal with clipping capacitance allows get sharp leading edge. As consequence the time resolution improved. The "payoff" for that is the introduction of amplifiers into the system. Among tested algorithms the best time resolution obtained with the signal fitting by "pulse function". And the best timing point chosen was close to the linear part of the signal appearance. The time resolution obtained with Na22 radioactive source corresponds to the linear approximation of the leading edge at the level of 10-20% of the signal amplitude. We continue to study different algorithms for data analysis and results will be reported later on.

VII. Conclusion

We performed an initial study of silicon photomultipliers as photodetectors with LYSO crystals for TOF-PET application. Our results were obtained with the DRS4 waveform digitizer and are consistent with previous results based on a constant fraction discrimination circuit (containing Ortec 9327, TAC567 and ADC114 modules). A few algorithms were investigated for time resolution analysis. The best time resolution obtained was with pulse function approximation of the signals. The time resolution improved with clipping of the SiPMs signals. We believe the obtained results can be improved. We consider this study as an initial step for design and production of our first TOF-PET module prototype [11].



Fig. 1. Schematic of the biasing and readout circuit used for the SiPM timing measurements.













Fig. 2. Traces of the STM and MPPC signals. The SiPMs illuminated by PiLas laser head. Traces are taken without clipping capacitance (a, b) and with it

(c, d) and with Ortec VT120C preamplifier (e, f, 10 pF capacitance).





Fig. 3. From the top, on right: 4 STM inside the Pomona box with printed circuit board. The support for radiators or crystals is on the left. The same for 1 STM is in the middle. The MPPC matrix, 4x4 cells inside Pomona box with printed circuit board and support for radiators or crystals are on the bottom.



Fig. 4. The schematic diagram of the readout with the PiLas laser. SiPm1, 2 – silicon photomultipliers, PiLas – picoseconds laser, DRS4 – waveform digitizer, PC – personal computer.



Fig. 5. Illustration for linear fit to the leading edge of the signal. a - attenuated NIM signal, b - signals of 2 STMs, illuminated by PiLas laser.



Fig. 6. Timing spectra, corresponding 25 photoelectrons. Time resolution (sigma) is 38.5 ps.



Fig. 7. External view of the TOF-PET setup.



Fig. 8. The schematic diagram of the readout with the DRS4.





Fig. 9. Traces of 2 STMs signals (a), and 2 MPPC signals (b) taken by Tektronix oscilloscope. 10 pF clipping capacitance. LYSO crystals irradiated by radioactive source Co60, Ortec VT120C preamplifiers; c - traces of 2 MPPCs signals taken

with DRS4. No clipping capacitance. No Ortec VT120C preamplifier, Co60 source.



Fig. 10. Example of time (top) and pulse height (bottom) distribution. The conditions: Hamamatsu MPPC, 3x3 mm²; LYSO crystal, 2x2x7 mm³; VT120C amplifier, Co60 radioactive source. 155 ps (sigma) time resolution.



Fig. 11. Example of time (top) and pulse height (bottom) distribution. The conditions: Hamamatsu MPPC, 3x3 mm²; LYSO cristal, 3x3x15 mm³; VT120C amplifier, Na22 radioactive source. 244 ps (sigma) time resolution, 10% (FWHM) photo peak resolution.



Fig. 12. Example of time (left) and pulse height (right) distribution. The conditions: STM, 3,5x3,5 mm²; LYSO crystal, 3x3x15 mm³;VT120C amplifier, Na22 radioactive source. 334 ps (sigma) time resolution, 13.5% (FWHM) photo peak resolution.

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